

- NAVSO P-3641A

Qualification Process

MQBL series converters are qualified to:

- MIL-STD-810F
 - consistent with RTCA/D0-160E
- SynQor's First Article Qualification

 consistent with MIL-STD-883F
- SynQor's Long-Term Storage Survivability Qualification
- SynQor's on-going life test

In-Line Manufacturing Process

- AS9100 and ISO 9001:2008 certified facility
- Full component traceability
- Temperature cycling
- Constant acceleration
- •24, 96, 160 hour burn-in
- Three level temperature screening

Primary referenced enableContinuous short circuit and overload protection

Fixed switching frequency

Output over-voltage shutdown

No opto-isolators

Clock synchronization

Remote sense

Input under-voltage and over-voltage shutdown

Specification Compliance

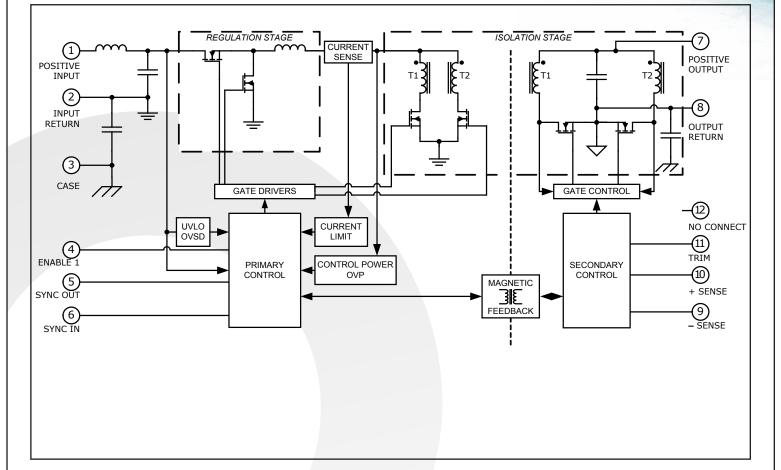
MQBL series converters (with MQHE filter) are designed to meet:

- MIL-HDBK-704-8 (A through F)
- RTCA/DO-160 Section 16, 17, 18
- MIL-STD-1275 (B, D) for $V_{IN} > 16V$
- DEF-STAN 61-5 (part 6)/(5, 6) for V_{IN} > 16V
- MIL-STD-461 (C, D, E, F)
- RTCA/DO-160(E, F, G) Section 22

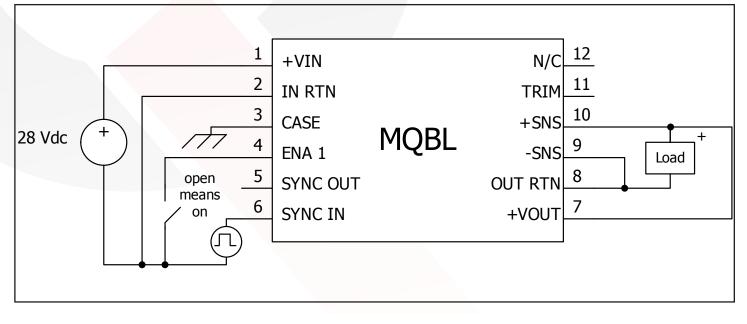
Phone 1-888-567-9596



BLOCK DIAGRAM



TYPICAL CONNECTION DIAGRAM



Page 2

Synce					MQBL-28-0 Output: 5V Current: 4A	
Technical Specific IQBL-28-05S ELECTRIC Parameter	CAL C Min.	CHAF Typ.	RACT Max.	ERIS Units	STICS Notes & Conditions Vin=28V dc ±5%, Iout=4A, CL=0µF, free running (see Note 9) unless otherwise specified	Group Subgro (see Note
BSOLUTE MAXIMUM RATINGS						
Non-Operating			60	V		
Operating			60	V	See Note 1	
Reverse Bias (Tcase = 125°C) Reverse Bias (Tcase = -55°C)			-0.8	V V		
colation Voltage (I/O to case, I to O)			-1.2	V		
Continuous	-500		500	V		
Transient (≤100µs)	-800		800	V	UD Curde Durdwetz, Care Nieters 2.0, 12	
perating Case Temperature orage Case Temperature	-55 -65		125 135	°C °C	HB Grade Products, See Notes 2 & 13	_
ad Temperature (20s)			300	°℃		
oltage at ENA1	-1.2		50	V		
NPUT CHARACTERISTICS	10	20	40	V	Cantinuaus	1.2
perating Input Voltage Range	16 16	28 28	40		Continuous Transient, 1s	1, 2, 4, 5,
put Under-Voltage Shutdown	10	20			See Note 3	., .,
Turn-On Voltage Threshold	14.75	15.50	16.00	V		1, 2,
Turn-Off Voltage Threshold	14.00	14.75	15.50			1, 2, 1, 2,
Shutdown Voltage Hysteresis put Over-Voltage Shutdown	0.65	0.85	1.05	V	See Note 3	1, Z,
Turn-Off Voltage Threshold	52.0	55.0	58.0	V		1, 2,
Turn-On Voltage Threshold	50.5	54.0	56.5	V		1, 2,
Shutdown Voltage Hysteresis	1.0	2.0	3.0	V		1, 2,
aximum Input Current o Load Input Current (operating)		40	1.7 80	A MA	Vin = 16V; Iout = 4A	1, 2,
isabled Input Current		10	15	mA	Vin = 16V, 28V, 50V; ENA	1, 2,
put Terminal Current Ripple (pk-pk)		20	35	mA	Bandwidth = 100kHz – 10MHz; see Figure 14	1, 2,
UTPUT CHARACTERISTICS	4.05	E 00	E 05			
utput Voltage Set Point (Tcase = 25°C) utput Voltage Set Point Over Temperature	4.95	5.00	5.05	V V	Vout at sense leads	2, 3
utput Voltage Line Regulation	-15	0	15	mV	"; Vin = 16V, 28V, 40V; Iout=4A	1, 2,
utput Voltage Load Regulation	-15	0	15	mV	"; Vout @ (Iout=0A) - Vout @ (Iout=4A)	1, 2,
otal Output Voltage Range	4.90	5.00	5.10	V	n	1, 2,
utput Over-Voltage Shutdown utput Voltage Ripple and Noise Peak to Peak	5.6	6.6 15	7.1	V mV	Pandwidth – 10MHz: CL – 11uE	See Not
perating Output Current Range	0	15	4	A	Bandwidth = 10MHz; CL=11µF	1, 2, 1, 2,
perating Output Power Range	0		20	Ŵ		1, 2,
utput DC Current-Limit Inception	4.2	5.1	6.1	A	See Note 4	1, 2,
ack-Drive Current Limit while Enabled		1.2	40	A		1, 2,
ack-Drive Current Limit while Disabled aximum Output Capacitance		5	40 2,000	mA µF		1, 2, See Not
YNAMIC CHARACTERISTICS	1		2,000	pi pi		
utput Voltage Deviation Load Transient					See Note 6	
For a Pos. Step Change in Load Current	-450	-300	450	mV	Total Iout step = 2A to 4A, 0.4A to 2A; $CL=11\mu F$	4, 5,
For a Neg. Step Change in Load Current utput Voltage Deviation Line Transient		300	450	mV	Vin step = 16V to 50V; $CL=11\mu$ F; see Note 7	4, 5,
For a Pos. Step Change in Line Voltage	-125		125	mV		4, 5,
For a Neg. Step Change in Line Voltage	-125		125	mV		4, 5,
Irn-On Transient			10		Vaut 0.5V/to 4.5V/t.5v/l. Desistive Load	4 5
Output Voltage Rise Time Output Voltage Overshoot		6	10	ms %	Vout = 0.5V to 4.5V; Full Resistive Load Resistive load	4, 5, See Not
Turn-On Delay, Rising Vin		5.5	8.0	ms	ENA = 5V; see Notes 8 & 10	4, 5,
Turn-On Delay, Rising ENA		3.0	6.0	ms	See Note 10	4, 5,
Restart Inhibit Time	10	100	150	ms	See Note 10	4, 5,
Short Circuit Start Time FFICIENCY	12	14	20	ms	Duration of pulse width, see Figure 16	4
ut = 4 A (16 Vin)	83	87		%		1, 2,
out = 2 A (16 Vin)	83	87		%		1, 2,
put = 4 A (28 Vin)	83	86		%		1, 2,
but = 2 A (28 Vin) but = 4 A (40 Vin)	80 81	85 85		% %		1, 2, 1, 2,
put = 2 A (40 Vin)	78	83		%		1, 2,
out = 4 A (50 Vin)	80	84		%		1, 2,
bad Fault Power Dissipation		1.3		l w	Sustained short circuit on output	

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Page 3

	Oí			AT T	MQBL-28-09 Output: 5V Current: 4A	
Technical Speci MQBL-28-05S ELECTRI Parameter		HAR	ACT Max.	ERIS Units	TICS (Continued) Notes & Conditions Vin=28V dc ±5%, Iout=4A, CL=0µF, free running (see Note 9) unless otherwise specified	Group A Subgroup (see Note 11)
ISOLATION CHARACTERISTICS	<u> </u>				Distantia sture atta	
Isolation Voltage	F00			V	Dielectric strength	1
Input RTN to Output RTN	500			V		
Any Input Pin to Case	500			•		1
Any Output Pin to Case	500			V		1
Isolation Resistance (in rtn to out rtn)	100			MΩ		1
Isolation Resistance (any pin to case)	100	22		MΩ		
Isolation Capacitance (in rtn to out rtn)		22		nF		1
FEATURE CHARACTERISTICS	E00	FFO	600			1 2 2
Switching Frequency (free running)	500	550	600	kHz		1, 2, 3
Synchronization Input	500		700	L.L.		1 2 2
Frequency Range	500 2.0		700	kHz V		1, 2, 3
Logic Level High				V		1, 2, 3
Logic Level Low	-0.5		0.8	V %		1, 2, 3
Duty Cycle	20		80	%		See Note 5
Synchronization Output Pull Down Current	20				VSYNC OUT = $0.8V$	See Note 5
	20		60	mA %		
Duty Cycle	40		60	%	Output connected to SYNC IN of another MQBL unit	See Note 5
Enable Control (ENA)			0.8	V		1 2 2
Off-State Voltage			0.8	-	Current dupin required to ensure module is off	1, 2, 3
Module Off Pulldown Current	80			μA V	Current drain required to ensure module is off	See Note 5
On-State Voltage	2		20		Imay draw from ain allowed with modulo still on	1, 2, 3 See Note 5
Module On Pin Leakage Current Pull-Up Voltage	3.2	4.0	20 4.8	μA V	Imax draw from pin allowed with module still on See Figure A	1, 2, 3
RELIABILITY CHARACTERISTICS	J.2	U	1.0	v		1, 2, 3
Calculated MTBF (MIL-STD-217F2)						
$GB @ Tcase = 70^{\circ}C$		2500		10 ³ Hrs.		
AIF @ Tcase = 70° C		190		10^{3} Hrs.		
WEIGHT CHARACTERISTICS		150		10 1113.		
Device Weight		35		q		
201.00 traight		55		9		

Electrical Characteristics Notes

1. Converter will undergo input over-voltage shutdown.

2. Derate output power for continuous operation per Figure 5.

3. High or low state of input voltage must persist for about 200µs to be acted on by the shutdown circuitry.

4. Current limit inception is defined as the point where the output voltage has dropped to 90% of its nominal value. See Current Limit discussion in Features Description section.

5. Parameter not tested but guaranteed to the limit specified.

6. Load current transition time \geq 10µs.

7. Line voltage transition time \geq 100µs.

8. Input voltage rise time \leq 250µs.

9. Operating the converter at a synchronization frequency above the free running frequency will cause the converter's efficiency to be slightly reduced

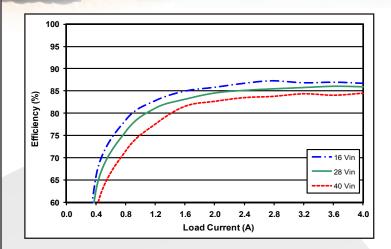
and it may also cause a slight reduction in the maximum output current/power available. For more information consult the factory. 10. After a disable or fault event, module is inhibited from restarting for 100ms. See Shut Down section of the Control Features description. 11. Only the ES and HB grade products are tested at three temperatures. The C grade products are tested at one temperature. Please refer to the

Construction and Environmental Stress Screening Options table for details.

 These derating curves apply for the ES and HB grade products. The C grade product has a maximum case temperature of 70°C.
 The specified operating case temperature for ES grade products is -45°C to 100°C. The specified operating case temperature for C grade products. is 0°C to 70°C.

MQBL-28-05S Output: 5V

Current: 4A



Technical Figures

Figure 1: Efficiency at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at Tcase=25°C.

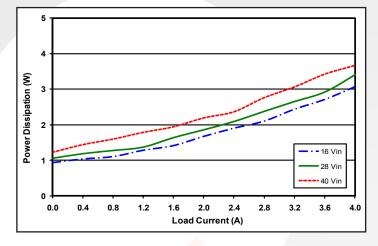


Figure 3: Power dissipation at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at Tcase=25°C.

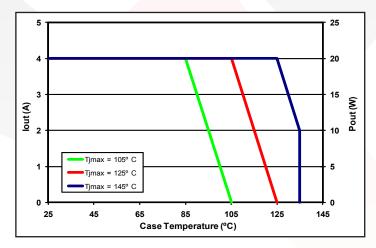


Figure 5: Output Current / Output Power derating curve as a function of Tcase and the Maximum desired power MOSFET junction temperature at Vin = 28V (see Note 12).

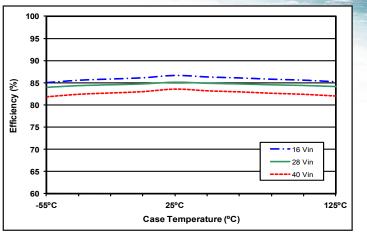


Figure 2: *Efficiency at nominal output voltage and 60% rated power vs. case temperature for input voltage of 16V, 28V, and 40V.*

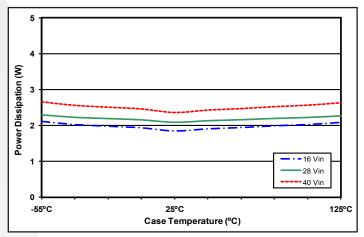


Figure 4: Power dissipation at nominal output voltage and 60% rated power vs. case temperature for input voltage of 16V, 28V, and 40V.

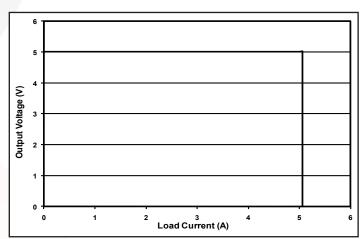


Figure 6: Output voltage vs. load current showing typical current limit curves at Vin = 28V.

Page 5

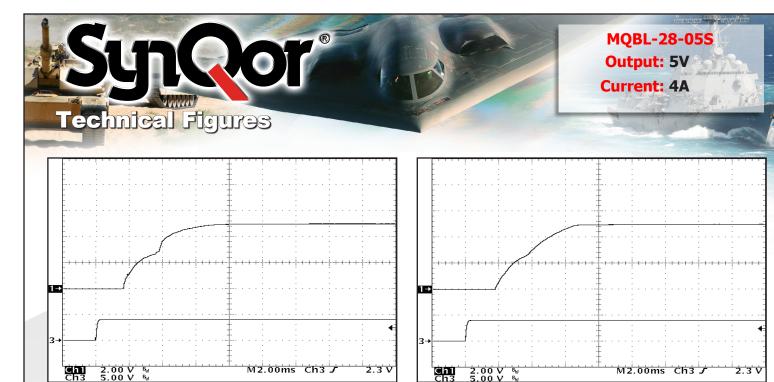


Figure 7: Turn-on transient at full resistive load and zero output capacitance initiated by ENA1. Input voltage pre-applied. Ch 1: Vout (2V/div). Ch 3: ENA1 (5V/div).

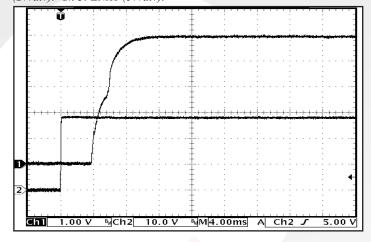


Figure 9: Turn-on transient at full resistive load and zero output capacitance initiated by Vin. ENA1 previously high. Ch 1: Vout (1V/ div). Ch 2: Vin (10V/div).

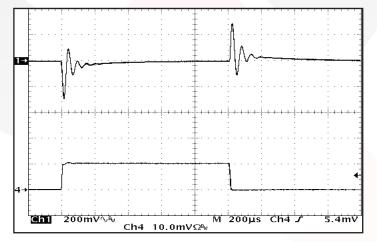


Figure 11: Output voltage response to step-change in load current 0%-50%-0% of Iout (max). Load cap: $1\mu F$ ceramic cap and $10\mu F$, $100m\Omega$ ESR tantalum cap. Ch 1: Vout (200mV/div). Ch 4: Iout (2A/div).

Figure 8: Turn-on transient at full resistive load and 1.5mF output capacitance initiated by ENA1. Input voltage pre-applied. Ch 1: Vout (2V/div). Ch 3: ENA1 (5V/div).

В,

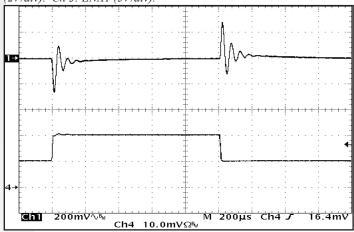


Figure 10: Output voltage response to step-change in load current 50%-100%-50% of lout (max). Load cap: $1\mu F$ ceramic cap and $10\mu F$, $100m\Omega$ ESR tantalum cap. Ch 1: Vout (200mV/div). Ch 4: Iout (2A/ div).

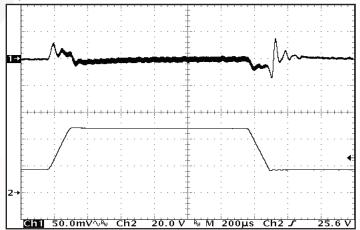
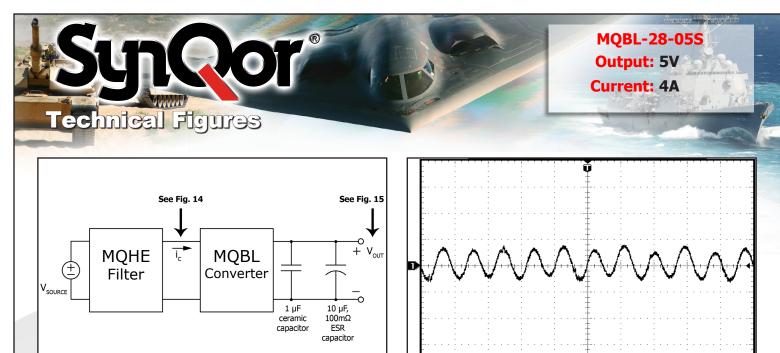
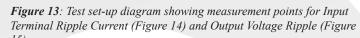


Figure 12: Output voltage response to step-change in input voltage (16V - 50V - 16V) in 150µS. Ch 1: Vout (50mV/div). Ch 2: Vin (20V/ div).





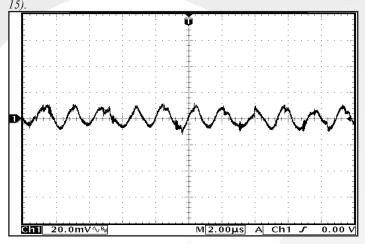


Figure 15: Output voltage ripple, Vout, at nominal input voltage and rated load current (20mV/div). Load capacitance: $1\mu F$ ceramic capacitor and $10\mu F$ tantalum capacitor. Bandwidth: 20MHz. See Figure

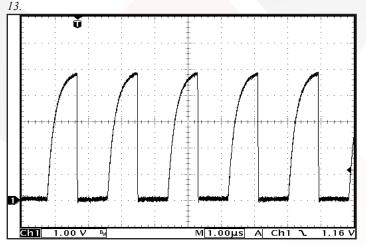
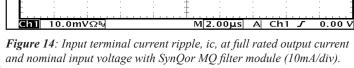


Figure 17: SYNC OUT vs. time, driving SYNC IN of a second SynQor MQBL converter. Ch1: SYNC OUT: (1V/div).



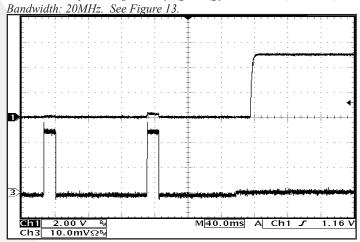
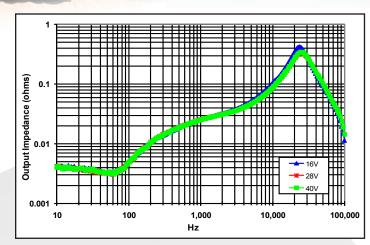


Figure 16: Rise of output voltage after the removal of a short circuit across the output terminals. Ch 1: Vout (2V/div). Ch 3: Iout (2A/div).

MQBL-28-05S Output: 5V

Current: 4A



Technical Figures

Figure 18: Magnitude of incremental output impedance (Zout = vout/ iout) for minimum, nominal, and maximum input voltage at full rated power.

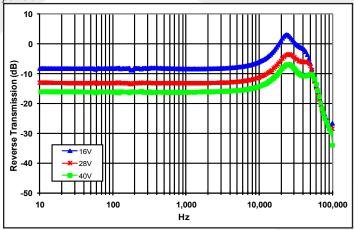


Figure 20: Magnitude of incremental reverse transmission (*RT* = *iin/ iout*) for minimum, nominal, and maximum input voltage at full rated power.

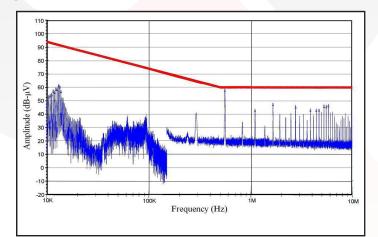


Figure 22: High frequency conducted emissions of standalone MQHL-28-05S, 5Vout module at 50W output, as measured with Method CE102. Limit line shown is the 'Basic Curve' for all applications with a 28V source.

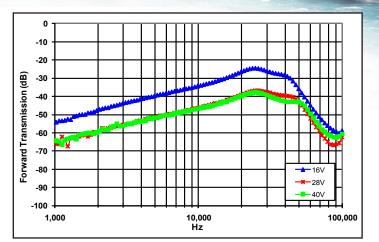


Figure 19: Magnitude of incremental forward transmission (FT = vout/ vin) for minimum, nominal, and maximum input voltage at full rated power.

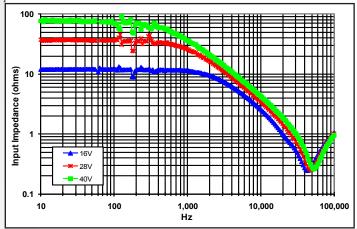


Figure 21: Magnitude of incremental input impedance (Zin = vin/iin) for minimum, nominal, and maximum input voltage at full rated power.

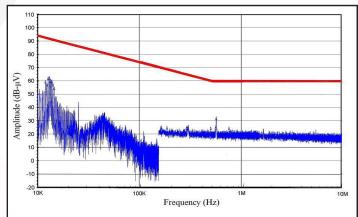


Figure 23: High frequency conducted emissions of MQHL-28-05S, 5Vout module at 50W output with MQHE-28-P filter, as measured with Method CE102. Limit line shown is the 'Basic Curve' for all applications with a 28V source.



BASIC OPERATION AND FEATURES

The MQBL DC/DC converter uses a two-stage power conversion topology. The first, or regulation, stage is a buck-converter that keeps the output voltage constant over variations in line, load, and temperature. The second, or isolation, stage uses transformers to provide the functions of input/output isolation and voltage transformation to achieve the output voltage required.

Both the regulation and the isolation stages switch at a fixed frequency for predictable EMI performance. The isolation stage switches at one half the frequency of the regulation stage, but due to the push-pull nature of this stage it creates a ripple at double its switching frequency. As a result, both the input and the output of the converter have a fundamental ripple frequency of about 550 kHz in the freerunning mode.

Rectification of the isolation stage's output is accomplished with synchronous rectifiers. These devices, which are MOSFETs with a very low resistance, dissipate far less energy than would Schottky diodes. This is the primary reason why the MQBL converters have such high efficiency, particularly at low output voltages.

Besides improving efficiency, the synchronous rectifiers permit operation down to zero load current. There is no longer a need for a minimum load, as is typical for converters that use diodes for rectification. The synchronous rectifiers actually permit a negative load current to flow back into the converter's output terminals if the load is a source of short or long term energy. The MQBL converters employ a "backdrive current limit" to keep this negative output terminal current small.

There is a control circuit in the MQBL converter that determines the conduction state of the power switches. It communicates across the isolation barrier through a magnetically coupled device. No opto-isolators are used.

An input under-voltage shutdown feature with hysteresis is provided, as well as an input over-voltage shutdown and an output over-voltage limit. There is also an output current limit that is nearly constant as the load impedance decreases (i.e., there is not fold-back or fold-forward characteristic to the output current under this condition). When a load fault is removed, the output voltage rises exponentially to its nominal value without an overshoot. If a load fault pulls the output voltage below about 60% of nominal, the converter will shut down to attempt to clear the load fault. After a short delay it will try to auto-restart.

The MQBL converter's control circuit does not implement an over-temperature shutdown.

The following sections describe the use and operation of additional control features provided by the MQBL converter.

CONTROL FEATURES

ENABLE: The MQBL converter has one enable pin, ENA1 (pin 4), which is referenced with respect to the converter's input return (pin 2). It must have a logic high level for the converter to be enabled; a logic low inhibits the converter.

The enable pin is internally pulled high so that an open connection will enable the converter. Figure A shows the equivalent circuit looking into the enable pin. It is TTL compatible and has hysteresis.

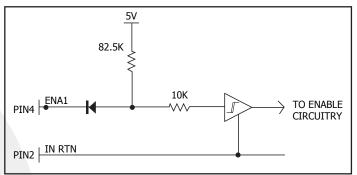


Figure A: Circuit diagram shown for reference only, actual circuit components may differ from values shown for equivalent circuit.

SHUT DOWN: The MQBL converter will shut down in response to only five conditions: ENA input low, VIN input below under-voltage shutdown threshold, VIN input above over-voltage shutdown threshold, output voltage below the output under-voltage threshold, and output voltage above the output over-voltage threshold. Following any shutdown event, there is a startup inhibit delay which will prevent the converter from restarting for approximately 100ms. After the 100ms delay elapses, if the enable inputs are high and the input voltage is within the operating range, the converter will restart. If the VIN input is brought down to nearly 0V and back into the operating range, there is no startup inhibit, and the output voltage will rise according to the "Turn-On Delay, Rising Vin" specification.

REMOTE SENSE: The purpose of the remote sense pins is to correct for the voltage drop along the conductors that connect the converter's output to the load. To achieve this goal, a separate conductor should be used to connect the +SENSE pin (pin 10) directly to the positive terminal of the load, as shown in the connection diagram on Page 2. Similarly, the –SENSE pin (pin 9) should be connected through a separate conductor to the return terminal of the load.

NOTE: Even if remote sensing of the load voltage is not desired, the +SENSE and the -SENSE pins must be connected to +Vout (pin 7) and OUTPUT RETURN (pin 8), respectively, to get proper regulation of the converter's output. If they are left open, the converter will have an output voltage that is approximately 200mV higher than its specified value.



Inside the converter, +SENSE is connected to +Vout with a 100Ω resistor and –SENSE is connected to OUTPUT RETURN with a 10Ω resistor.

It is also important to note that when remote sense is used, the voltage across the converter's output terminals (pins 7 and 8) will be higher than the converter's nominal output voltage due to resistive drops along the connecting wires. This higher voltage at the terminals produces a greater voltage stress on the converter's internal components and may cause the converter to fail to deliver the desired output voltage at the low end of the input voltage range at the higher end of the load current and temperature range. Please consult the factory for details.

SYNCHRONIZATION: The MQBL converter's switching frequency can be synchronized to an external frequency source that is in the 500 kHz to 700 kHz range. A pulse train at the desired frequency should be applied to the SYNC IN pin (pin 6) with respect to the INPUT RETURN (pin 2). This pulse train should have a duty cycle in the 20% to 80% range. Its low value should be below 0.8V to be guaranteed to be interpreted as a logic low, and its high value should be above 2.0V to be guaranteed to be interpreted as a logic high. The transition time between the two states should be less than 300ns.

If the MQBL converter is not to be synchronized, the SYNC IN pin should be left open circuit. The converter will then operate in its free-running mode at a frequency of approximately 550 kHz.

If, due to a fault, the SYNC IN pin is held in either a logic low or logic high state continuously, or the SYNC IN frequency is outside the 500-700 kHz range, the MQBL converter will revert to its free-running frequency.

The MQBL converter also has a SYNC OUT pin (pin 5). This output can be used to drive the SYNC IN pins of as many as ten (10) other MQBL converters. The pulse train coming out of SYNC OUT has a duty cycle of 50% and a frequency that matches the switching frequency of the converter with which it is associated. This frequency is either the free-running frequency if there is no valid synchronization signal at the SYNC IN pin, or the synchronization frequency if there is.

The synchronization feature is entirely compatible with that of SynQor's MQFL family of converters.

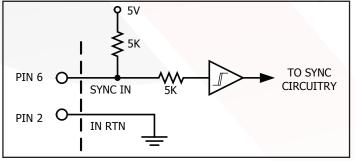


Figure B: Equivalent circuit looking into the SYNC IN pin with respect to the IN RTN (input return) pin.

Figure B shows the equivalent circuit looking into the SYNC IN pin and Figure C shows the equivalent circuit looking into the SYNC OUT pin.

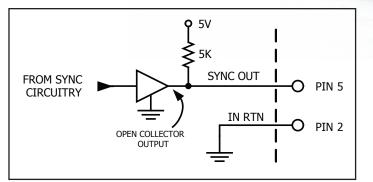


Figure C: Equivalent circuit looking into SYNC OUT pin with respect to the IN RTN (input return) pin.

OUTPUT VOLTAGE TRIM: the TRIM pin (pin 11) can adjust the MQBL converter's output voltage $\pm 10\%$ around its nominal value.

To trim the output voltage above its nominal value, connect an external resistor from the TRIM pin to the –SENSE pin as shown in Figure D. The value of this trim up resistor should be chosen according to the following equation or from Figure E:

Rtrim
$$up(\Omega) = \frac{6000\Omega * Vnom}{Vout - Vnom} - 30000\Omega$$

where:

Vnom = the converter's nominal output voltage, Vout = the desired output voltage (greater than Vnom), and Rtrim up is in Ohms.

As the output voltage is trimmed up, it produces a greater voltage stress on the converter's internal components and may cause the converter to fail to deliver the desired output voltage at the low end of the input voltage range at the higher end of the load current and temperature range. Please consult the factory for details. To trim the output voltage below its nominal value, connect an external resistor between the TRIM pin and the +SENSE pin. The value of this trim down resistor should be chosen according to the following equation or from Figure E:

$$\begin{array}{rl} \mbox{Rtrim down}(\Omega) = & \underline{15100\Omega^* \mbox{Vout} - 6000\Omega^* \mbox{Vnom}} \\ \mbox{Vnom} & - 30000\Omega \end{array}$$

where:

Vnom = the converter's nominal output voltage, Vout = the desired output voltage (less than Vnom), and

Rtrim down is in Ohms.

Factory trimmed converters are available by request.



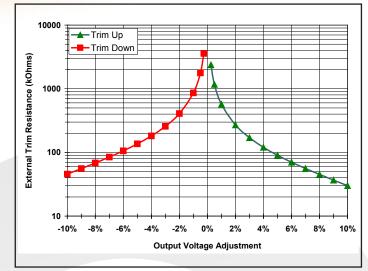


Figure E: Trim up and Trim down as a function of external trim resistance.

INPUT UNDER-VOLTAGE SHUTDOWN: The MQBL converter has an under-voltage shutdown feature that ensures the converter will be off if the input voltage is too low. The input voltage turn-on threshold is higher than the turn-off threshold. In addition, the MQBL converter will not respond to a state of the input voltage unless it has remained in that state for more than about 200µs. This hysteresis and the delay ensure proper operation when the source impedance is high or in a noisy environment.

INPUT OVER-VOLTAGE SHUTDOWN: The MQBL converter also has an over-voltage feature that ensures the converter will be off if the input voltage is too high. It also has a hysteresis and time delay to ensure proper operation.

OUTPUT OVER-VOLTAGE SHUTDOWN: The MQBL converter will shut down if the voltage at its power output pins ever exceeds about 130% of the nominal value. The

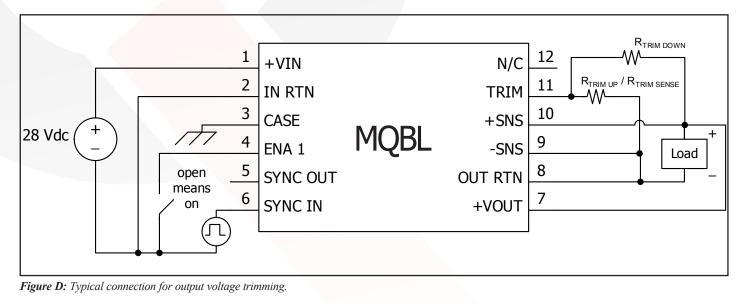
shutdown threshold does not change with output trim or sense drops; excessive trim-up or output wiring drops may cause an output over-voltage shutdown event. After a startup inhibit delay, the converter will attempt to restart.

OUTPUT UNDER-VOLTAGE SHUTDOWN: The MQBL converter will also shut down if the voltage at its power output pins ever dips below 60% of the nominal value for more than a few milliseconds. Output voltage reduction due to output current overload (current limit) is the most common trigger for this shutdown. The shutdown threshold does not change with output trim but at only 10%, trimdown should not trigger this event. After a startup inhibit delay, the converter will attempt to restart. This shutdown is disabled during startup.

BACK-DRIVE CURRENT LIMIT: Converters that use MOSFETs as synchronous rectifiers are capable of drawing a negative current from the load if the load is a source of short- or long-term energy. This negative current is referred to as a "back-drive current".

Conditions where back-drive current might occur include paralleled converters that do not employ current sharing. It can also occur when converters having different output voltages are connected together through either explicit or parasitic diodes that, while normally off, become conductive during startup or shutdown. Finally, some loads, such as motors, can return energy to their power rail. Even a load capacitor is a source of back-drive energy for some period of time during a shutdown transient.

To avoid any problems that might arise due to back-drive current, the MQBL converters limit the negative current that the converter can draw from its output terminals. The threshold for this back-drive current limit is placed sufficiently below zero so that the converter may operate properly down to zero load, but its absolute value (see the Electrical Characteristics page) is small compared to the converter's rated output current.



MQBL-28-05S Output: 5V Current: 4A

Application Section

CURRENT LIMIT: In the event of excess load, the MQBL converter will quickly reduce its output voltage to keep the load current within safe limits (see Figure 6). If the overload persists for more than 14 milliseconds, the converter will shut off, wait a restart delay, and then automatically attempt to re-start. The timeout is internally implemented with an integrator: counting up whenever current limit is active, and counting down at 1/5th the rate whenever current limit becomes inactive. In this way a series of short-duration overloads will not cause the converter to shut down, while it will shut down in response to sustained overloads.

THERMAL CONSIDERATIONS: Figure 5 shows the suggested Power Derating Curves for this converter as a function of the case temperature and the maximum desired power MOSFET junction temperature. All other components within the converter are cooler than its hottest MOSFET, which at full power is no more than 20°C higher than the case temperature directly below this MOSFET.

The Mil-HDBK-1547A component derating guideline calls for a maximum component temperature of 105°C. Figure 5 therefore has one power derating curve that ensures this limit is maintained. It has been SynQor's extensive experience that reliable long-term converter operation can be achieved with a maximum component temperature of 125°C. In extreme cases, a maximum temperature of 145°C is permissible, but not recommended for long-term operation where high reliability is required. Derating curves for these higher temperature limits are also included in Figure 5. The maximum case temperature at which the converter should be operated is 135°C. When the converter is mounted on a metal plate, the plate will help to make the converter's case bottom a uniform temperature. How well it does so depends on the thickness of the plate and on the thermal conductance of the interface layer (e.g. thermal grease, thermal pad, etc.) between the case and the plate. Unless this is done very well, it is important not to mistake the plate's temperature for the maximum case temperature. It is easy for them to be as much as 5-10°C different at full power and at high temperatures. It is suggested that a thermocouple be attached directly to the converter's case through a small hole in the plate when investigating how hot the converter is getting. Care must also be made to ensure that there is not a large thermal resistance between the thermocouple and the case due to whatever adhesive might be used to hold the thermocouple in place.

INPUT SYSTEM INSTABILITY: This condition can occur because any dc-dc converter appears incrementally as a negative resistance load. A detailed application note titled "Input System Instability" is available on the SynQor website which provides an understanding of why this instability arises, and shows the preferred solution for correcting it.

MQBL-28-05S Output: 5V

Current: 4A

Screening	Consistent with MIL-STD-883F	C-Grade (specified from (0 °C to +70 °C)	ES-Grade specified from (-45 °C to +100 °C)	HB-Grade specified from -55 °C to +125 °C
Element Evaluation		No	Yes	Yes
Internal Visual	*	Yes	Yes	Yes
Temperature Cycle	Method 1010	No	Condition B (-55 °C to +125 °C)	Condition C (-65 °C to +150 °C)
Constant Acceleration	Method 2001 (Y1 Direction)	No	500g	Condition A (5000g)
Burn-in	Method 1015	24 Hrs @ +125 ℃	96 Hrs @ +125 ℃	160 Hrs @ +125 °C
Final Electrical Test	Method 5005 (Group A)	+25 °C	-45, +25, +100 °C	-55, +25, +125 °C
Mechanical Seal, Thermal, and Coating Process		Full QorSeal	Full QorSeal	Full QorSeal
External Visual	2009	*	Yes	Yes
Construction Process		QorSeal	QorSeal	QorSeal
	* Per IPC-A	-610 Class 3		

R

Stress Screening

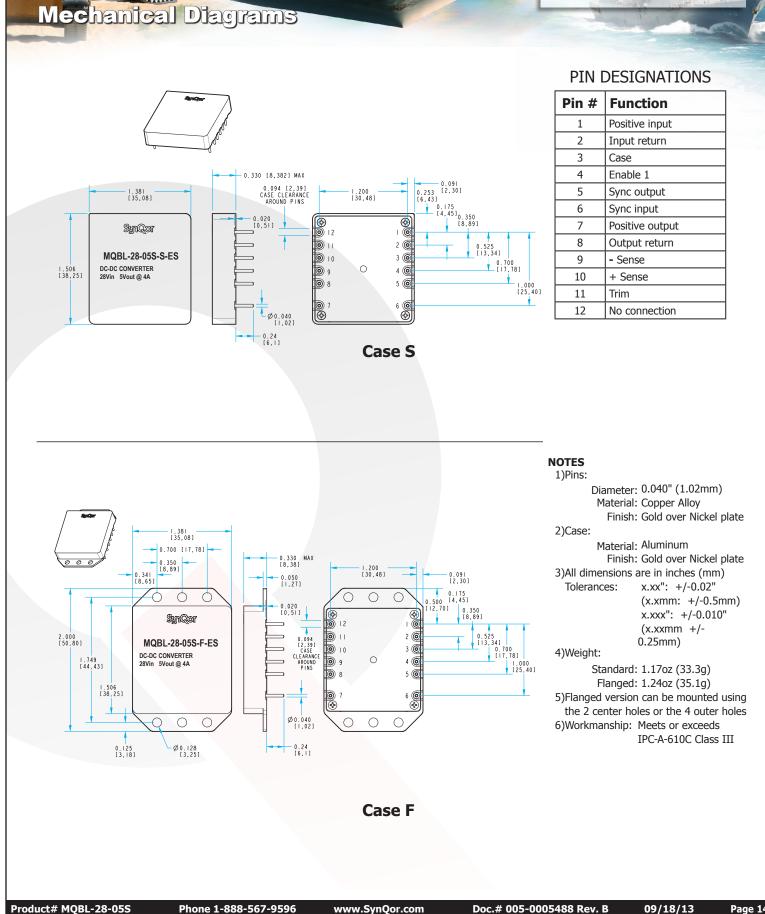
MilQor converters and filters are offered in three variations of environmental stress screening options. All MilQor converters use SynQor's proprietary QorSealTM Hi-Rel assembly process that includes a Parylene-C coating of the circuit, a high performance thermal compound filler, and a nickel barrier gold plated aluminum case. Each successively higher grade has more stringent mechanical and electrical testing, as well as a longer burn-in cycle. The ES-and HB-Grades are also constructed of components that have been procured through an element evaluation process that pre-qualifies each new batch of devices.

Product# MQBL-28-05S

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MQBL-28-05S Output: 5V

Current: 4A



09/18/13



MilQor Converter FAMILY MATRIX

The tables below show the array of MilQor converters available. When ordering SynQor converters, please ensure that you use the complete part number according to the table in the last page. Contact the factory for other requirements.

	Single Output											Dua	l Outp	ut †		
	1.5V	1.8V	2.5V	3.3V	5V	6V	7.5V	9V	12V	15V	28V		5V	12V	15V	
	(1R5S)	(1R8S)	(2R5S)	(3R3S)	(05 S)	(06S)	(7R5S)	(09S)	(12S)	(15S)	(28S)		(05D)	(12D)	(15D)	
MQHL-28																
16-40Vin Cont.	20A	20A	20A	15A	10A	8A	6.6A	5.5A	4A	3.3A	1.8A		10A	4A	3.3A	
16-50Vin 1s Trans.*	20A	20A	20A	IJA	IUA	оA	0.0A	5.5A	44	5.5A	1.0A		Total	Total	Total	
Absolute Max Vin = 60V																
MQHL-28E																
16-70Vin Cont.	20A	20A	20A	15A	10A	8A	6.6A	5.5A	4A	3.3A	1.8A		10A	4A	3.3A	
16-80Vin 1s Trans.*	20A	20A	204	IJA	IUA	OA	0.0A	5.5A	44	5.5A	1.0A		Total	Total	Total	
Absolute Max Vin =100V																
MQHR-28												Γ				
16-40Vin Cont.	10A	10A	10A	7.5A	5A	4A	3.3A	2.75A	2A	1.65A	0.9A		5A	2A	1.65A	
16-50Vin 1s Trans.*	IUA	IUA	IUA	7.5A	5A	4A	3.3A	2.75A		1.05A	0.9A		Total	Total	Total	
Absolute Max Vin = 60V																
MQHR-28E												Γ				
16-70Vin Cont.	104	104	10.4	7 5 4	5A	4.0	3.3A	2 75 4	24	1 65 4			5A	2A	1.65A	
16-80Vin 1s Trans.*	10A	10A	10A 10A	IUA	7.5A	5A	4A	3.3A	2.75A	2.75A 2A	1.65A 0.9A	U.9A		Total	Total	Total
Absolute Max Vin = 100V												L				

	Single Output										Dual Output [†]			
	1.5V	1.8V	2.5V	3.3V	5V	6V	7.5V	9V	12V	15V	28V	5V	12V	15V
	(1R5S)	(1R8S)	(2R5S)	(3R3S)	(05S)	(06S)	(7 R 5S)	(09S)	(12S)	(15S)	(28S)	(05D)	(12D)	(15D)
MQBL-28														
16-40Vin Cont.	8A	8A	8A	6A	4A	3.3A	2.6A	2.2A	1.6A	1.3A	0.7A	4A	1.6A	1.3A
16-50Vin 1s Trans.*												Total	Total	Total
Absolute Max Vin = 60V														
MQBL-28E														
16-70Vin Cont.	8A	8A	8A	6A	4A	3.3A	2.6A	2.2A	1.6A	1.3A	0.7A	4A	1.6A	1.3A
16-80Vin 1s Trans.*		0A		UA	4A	5.5A	2.0A	2.2A	1.0A	1.3A	0.7A	Total	Total	Total
Absolute Max Vin =100V														

Check with factory for availability. †80% of total output current available on any one output. *Converters may be operated at the highest transient input voltage, but some component electrical and thermal stresses would be beyond MIL-HDBK-1547A guidelines.



PART NUMBERING SYSTEM

The part numbering system for SynQor's MilQor DC-DC converters follows the format shown in the table below.

Not all combinations make valid part numbers, please contact SynQor for availability. See the Product Summary web page for more options.

Model	Input	Output V	/oltage(s)	Package Outline/	Screening		
Name	Voltage Range	Single Output	Dual Output	Pin Configuration	Grade		
MQBL	28 28E	1R5S 1R8S 2R5S 3R3S 05S 06S 7R5S 09S 12S 15S 28S	05D 12D 15D	S F	C ES HB		

Example: MQBL-28-05S-F-ES

APPLICATION NOTES

A variety of application notes and technical white papers can be downloaded in pdf format from the SynQor website.

PATENTS

SynQor holds the following U.S. patents, one or more of which apply to each product listed in this document. Additional patent applications may be pending or filed in the future.

5,999,417	6,222,742	6,545,890	6,577,109	6,594,159	6,731,520
6,894,468	6,896,526	6,927,987	7,050,309	7,072,190	7,085,146
7,119,524	7,269,034	7,272,021	7,272,023	7,558,083	7,564,702
7,765,687	7,787,261	8,023,290	8,149,597		

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Warranty

SynQor offers a two (2) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

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